# **Design of Modified Booth's Encoder Using SPST technique**

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Abstract: This paper presents the design and implementation of signed-unsigned Modified Booth Encoding (SUMBE) multiplier. The present Modified Booth Encoding (MBE) multiplier and the Baugh-Wooley multiplier perform multiplication operation on signed numbers only. Therefore, this paper presents the design and implementation of SUMBE multiplier. The modified Booth Encoder circuit generates half the partial products in parallel. By extendingsign bit of the operands and generating an additional partial product the SUMBE multiplier is obtained. The Carry Save Adder (CSA) tree andthe final Carry Look ahead (CLA) adder used to speed up the multiplier operation. Since signed and unsigned multiplication operation is performed by the same multiplier unit the required hardware and the chip area reduces and this in turn reduces power dissipation and cost of a system. The proposed radix-2 modified Booth algorithm MAC with SPST gives a factor of 5 lessdelay and 7% less power consumption as compared to array MAC.

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### **1. Introduction**

### **1.1Common Features of Multipliers:**

### **1.1.1 Counterflow Organization:**

A novel multiplier organization is introduced, in which the data bits flow in one direction, and the Booth commands [1] are piggybacked on the acknowledgments flowing in the opposite direction.

### 1.1.2 Merged Arithmetic/Shifter Unit:

An architectural optimization is introduced that merges the arithmetic operations and the shift operation into the same function unit, thereby obtaining significant improvement in area, energy and speed[1].

### 1.1.3 Overlapped Execution:

The entire design is pipelined at the bitlevel, which allows overlapped execution of Proceedings of multiple iterations of the Booth algorithm, including across successive multiplications. As a result, both the cycle time per Booth iteration, as well as the overall cycle time per multiplication are significantly improved[2].

### 1.1.4 Modular Design:

The design is quite modular, which allows the implementation to be scaled to arbitrary operand widths[2] without the need for gate resizing, and without incurring any overhead on iteration time.

### 1.1.5 Precision-Energy Trade-Off:

Finally, the architecture can be easily modified to allow dynamic specification of operand widths, i.e., successive operations of a given multiplier implementation could operate upon different word length[4].

A new architecture of multiplier and accumulator (MAC) for high-speed arithmetic. By combining multiplication with accumulation **was** and devising a hybrid type of carry save adder (CSA), the performance was improved. Since the accumulator that has the largest delay in MAC was merged into CSA, the overall performance elevated. The proposing method CSA tree uses 1's-complementbased radix-2 modified Booth's algorithm (MBA) and has the modified array for the sign extension in order to increase the bit density of the operands. The proposed MAC showed the superior properties to the standard design in many ways and performance twice as much as the previous researchin the similar clock frequency. We expect that the proposed MAC[6] can be adapted to various fields requiring high performance such as the signal processing areas. The advanced digital processors now have fast bit-parallel multipliers embedded in them. Multipliers for unsigned numbers are designed using dizzying array of ways with each method having its own advantages and tradeoffs. Inrecent years, high-speed multipliers play an important role while designing any architecture and researchers are still working on many factors to increase the speed of operation of these basic elements<sup>[7]</sup>. Algorithms for designing high-speed multipliers have been modified and developed for better efficiency.

The increased complexity of various applications, demands not only faster multiplier chips but also smarter and efficient multiplying algorithms that can be implemented in the chips[8]. It is up to the need of the hour and the application on to which the multiplier is implemented and what tradeoffs need to be considered. Generally, the efficiency of the multipliers is classified based on the variation in speed, area and configuration. Due to rapidly growing system-on-chip industry, not only the faster units but also smaller area and less power has become a major concern for designing very large scale integration (VLSI) circuits[11]. Digital circuits make use of digital arithmetic's. Among various arithmetic operations, multiplication is one of the fundamental operation used and is being performed by an added.

There are many ways to build a multiplier each providing trade-off between delays and other characteristics[15], such as area and energy dissipation The objective of a good multiplier and accumulator (MAC)[24] is to provide a physically compact, good speed and low power consuming chip. To save significant power consumption of a VLSI design[16], it is a good direction to reduce its dynamic power that is the major part of total power dissipation. This paper proposes a high speed MAC adopting the new Spurious Power Suppression Technique (SPST) implementing approach This multiplier and accumulator is designed by equipping the SPST on a modified Booth encoder which is controlled by a detection unit using an AND gate. The modified booth encoder will reduce the number of partial products generated by a factor of 2. The SPST adder will avoid the unwanted addition and thus minimize the switching power dissipation.[17]

# 2. Related Work

The number and variety of products that include some form of digital signal processing has grown dramatically over the last years[25]. DSP has become a key component in many consumers. communications, medical, and industrial products. These products use a variety of hardware approaches[26] to implement DSP, ranging from the use of off-the-shelf microprocessors to fieldprogrammable gate arrays (FPGAs) to custom integrated circuits (ICs). Programmable "DSP processors," a class of microprocessors optimized for DSP[12], are a popular solution for several reasons. In comparison to fixed-function solutions, they have the advantage of potentially being reprogrammed in the field, allowing product upgrades or fixes. They are often more costeffective (and less risky) than custom hardware, particularly for low-volume applications, where the development cost of custom ICs[13] may be prohibitive. And in comparison to other types of microprocessors, DSP processors often have an advantage in terms of speed, cost, and energy efficiency [1].

. There are many implementation media available for signal processing. These implementations vary in terms of programmability from fixedfunctionality hardware like ASIC's [9] to fully programmable like general-purpose processors. The emergence of the new architecture[5], which offers the same computational attributes as fixedfunctionality architectures in a package that can be customized in the field, is driven by a need for realtime performance within the given operational parameters of a target system[14] and a need to adapt to changing data sets, computing conditions, and execution environments of DSP applications. In this paper we used three main ideas VHDL, architecture pipelining, and implementation of FPGAs. More details on FPGAs can be found in [2]This design has been simulated using the modelsim software then implemented on FPGA.

X <sub>i+1</sub>	Xi	Xi-1	Action
0	0	0	0  imes Y
0	0	1	1 × Y
0	1	0	1 × Y
0	1	1	$2 \times Y$
1	0	0	-2 × Y
1	0	1	-1 × Y
1	1	0	-1 × Y
1	1	1	0 × Y

Table 1. Recoding of bits using Modified Booths Encoder

### 3. Modified Booth Encoder

In order to achieve high-speed, multiplication algorithms using parallel counters, such as the modified Booth algorithm[18] has been proposed, and some multipliers based on the algorithms have been implemented for practical use. This type of multiplier operates much faster than an array multiplier for longer operands because its computation time is proportional to the logarithm of the word length of operands[19]. Booth multiplication is a technique that allows for To Booth recode the multiplier term,[20] we consider the bits in blocks of three, such that each block overlaps the previous block by one bit. Grouping starts from the LSB, and the first block only uses two bits of the multiplier. Figure 1 shows the grouping of bits from the multiplier term for use in modified booth encoding.

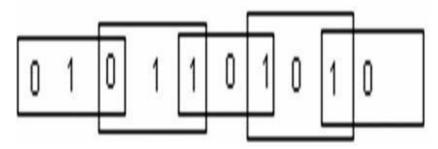


Figure 1.Recoding of bits

								0	0	0	0	0	1	0	(	)	0	8	
							х	0	0	0	0	1	0	1	(	)	0	20	
C	) (	)	0	 0	0	0	 0	0	0	0	0	0	0	0	(	)	0	0 × Y	
0	) (	)	0	0	0	0	0	0	0	0	1	0	0	0				$1 \times Y$	
+ (	) (	)	0	0	0	0	0	0	1	0	0	0						$1 \times Y$	
0	) (	)	0	0	0	0	0	0	0	0								$0 \times Y$	
0	) (	)	0	0	0	0	0	0										$0 \times Y$	
0	) (	)	0	0	0	0	0	0	1	0	1	0	0	0	(	)	0	160	

Figure 2 Example for Modified Booth Encoder

## 4. Proposed Low Power High Performance Multiplier and Accumulator

The proposed high speed low power multiplier is designed by equipping the SPST on a tree multiplier. There are two distinguishing design considerations in designing the proposed multiplier as listed in the following: Applying the SPST on the Modified Booth Encoder Figure 8 shows a computing example of Booth multiplying two numbers "2AC9" and "006A". The shadow denotes that the numbers in this part of Booth multiplication are all zero so that this part of the computations can be neglected. Saving those computations can significantly reduce the power consumption caused by the transient signals. According to the analysis of the multiplication shown in figure 8, we propose the SPST-equipped modified-Booth encoder, which is controlled by a detection unit. The detection unit has one of the two operands as its input to decide whether the Booth encoder calculates redundant computations as shown in Figure 9. The latches can, respectively, freeze the inputs of MUX-4 to MUX-7 or only those of MUX-6 to MUX-7 when the PP4 to PP7 or the PP6 to PP7 are zero; to reduce the transition power dissipation. Figure 10. Shows the booth partial product generation circuit. It includes AND/OR/ EX-OR logic. The former SPST has been discussed in [9] and [10]. Figure 4 shows the five cases of a 16-bit addition in which the spurious switching activities occur. The 1st case illustrates a transient state in which the spurious transitions of carry signals occur in the MSP though the final result of the MSP are unchanged[21]. The 2nd and the 3rd cases describe he situations of one negative operand adding another positive operand without and with carry from LSP, respectively. Moreover, the 4th and the 5th cases respectively demonstrate the addition of two negative operands without and with carry-in from LSP[22]. In those cases, the results of the MSP are predictable Therefore the computations in the MSP are useless and can be neglected. The data are separated into the Most Significant Part (MSP) and the Least Significant Part (LSP)[23].

The SPST uses a detection logic circuit to detect the effective data range of arithmetic units, e.g., adders or multiplier[24]s. When a portion of data does not affect the final computing results, the data controlling circuits of the SPST latch this portion to avoid useless data transitions occurring inside the arithmetic units. Besides, there is a data asserting control realized by using registers to further filter out the useless spurious signals of arithmetic unit every time when the latched portion is being turned on. This asserting control brings evident power reduction. Figure 5 shows the design of low power adder/subtract with SPST.

The adder /subtract is divided into two parts, the most significant part (MSP) and the least significant part (LSP). The MSP of the original adder/subtract is modified to include detection logic circuits, data controlling circuits, sign extension circuits, logics for calculating carry in and carry out signals. The most important part of this study is the design of the control signal asserting circuits, denoted as asserting circuits in Figure 2. Although this asserting circuit brings evident power reduction, it may induce additional delay. There are two implementing approaches for the control signal assertion circuits. The first implementing approach of control signal assertion circuit is using registers. This is illustrated in Figure 6. The three output signals of the detection logic are close, Carr ctrl, sign. The restriction that must be greater than to guarantee the registers from latching the wrong values of control usually decreases the overall speed of the applied designs

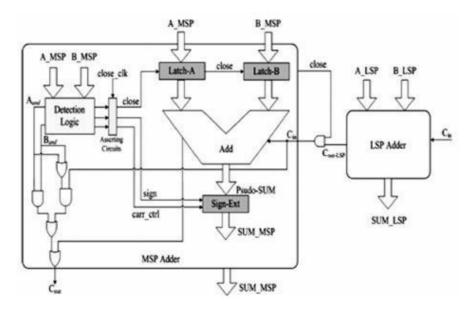


Figure 4.Booths Encoder using LSP and MSPadder

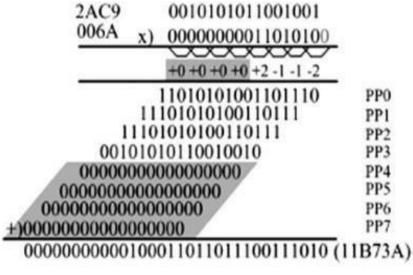


Figure 5: Illustration of Multiplication using Modified Booth Encoding

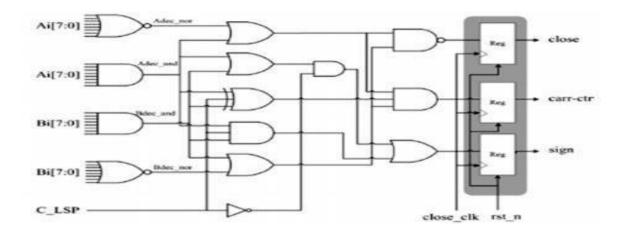


Figure 6 Booth Partial Product Selector Logic

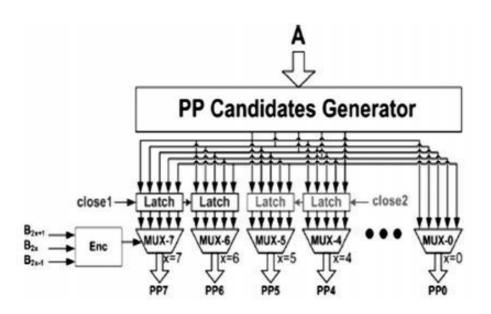


Figure 7.: SPST Equipped Modified Booth encoder

A. Applying the SPST on the Compression Tree The proposed SPST -equipped multiplier is

illustrated in figure 11. The PP generator generates five candidates of the partial products, i.e.,  $\{-2A, -A, 0, A, 2A\}$ . These are then selected according to the Booth encoding results of the operand *B*. When the operand besides the Booth encoded one has a small absolute value, there are opportunities to reduce the spurious power dissipate Radix-2 modified booth MAC with SPST performs both multiplication and accumulation. Multiplication result is obtained by multiplying multiplicand and multiplier. This multiplication with SPST module is shown in figure 8.

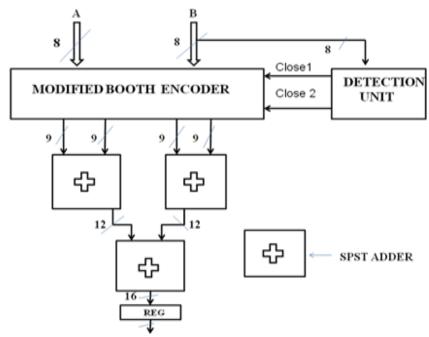


Figure 8 MAC with SPST Module

The schematic of MAC unit with SPST Module is obtained using the RTL schematic by Xilinx tool. The RTL schematic of the MAC with SPST module consist of a 16 bit input and a 32bit Output.

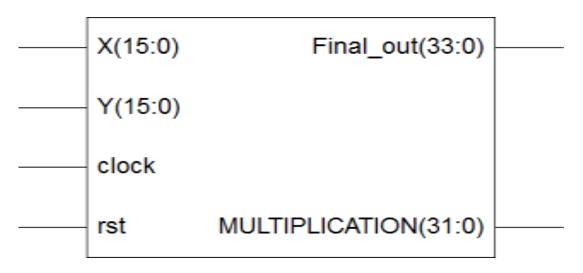


Figure 9. RTL Schematic of MAC Unit.

### **5. Results and Discussion**

The MBE code is written in verilog, synthesized and simulated using Xilinx and modelsim tools. The netlist generated during the execution of RC compiler tool from cadence is given as an input to the physical design. The portioning process and floor planning is done by using cadence assura tool. The W/L ratio is fixed at 0.33 for all the horizontal and vertical pads and rings and also the stripes. Next power planning is done. After that placement of the pads is done in such a way that the aspect ratio is minimized .Next routing is done for interconnection between the wires .Finally GSSII IS done and the ASCII code is generated, which will be further taken as an input to the fabrication process.

The results of FPGA and ASIC are as shown in Figures 10 and 11.The fig 10 describes the waveform i.e the simulation result for an 8x8 bit MBE. The final 32 bit output is obtained and it is divided into y0,y1,y2 and y3.The physical design output is obtained by using the assura tool from cadence and the netlist is generated from the RC compiler tool from cadence. Finally we compare FPGA and ASIC Metholologies.

Now: 1000 ns		200	4(	0	1	600	1	60.0
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olk 👔	0							
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🖬 🚮 p1_out(8:0)	9				9	15000001000		
🖬 🚮 p2_out[8:0]	9				9	15000001000		
🖬 🚮 p3_out(8:0)	9				9	100000000		
<b>31</b> s0	0							
<b>ði</b> s1	0							
<b>6</b> \$2	0							
<b>ð</b> ] s3	0							
<b>31</b> n0	0							
<b>31</b> n1	0							

Figure10.Simulation results for MBE using Modelsim

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Figure 11. Final Routing Output

also less as compared to it. It is completely dependon the Algorithm used in both Multipliers.

VENDOR	FPGA	ASIC
AREA	62%	16.3%
POWER DISSIPATION	123mw	32763.567nw
DELAY	1.901ns	7950ps
DELAY	1.901115	7950ps

Table 2 Comparison of Modified booth encoder using FPGA and ASIC Hence ASIC is preferred when compared to FPGA.

## 6. Conclusion and Scope for Future Work

Radix-2 Booth Multiplier is implemented here; the complete process of the implementation is giving higher speed of operation. The four cycle of shifting process including addition and subtraction is available. Now at the same time RTL Schematic generated here is giving the comfortable execution of it. This RTL Schematic can be implemented in FPGA CPLD kit that will give the proper Output. Now this RTL Schematic of Radix-2 Booth Multiplier is compared with implemented RTL Radix-4 Encoder Booth Multiplier. The Speed and Circuit Complexity is compared, Radix-4 Booth Multiplier is giving higher speed as compared to Radix-2 Booth Multiplier and Circuit Complexity is The MAC process is coded with VHDL and synthesized using Xilinx ISE 6.2i. The MAC process is implemented using xc3s1000-5fg456 FPGA Xilinx device. The synthesis results of the MAC unit have been calculated as can be seen in Table2. Here, same FPGA device (part number & speed grade) with the same design constraints implied for the synthesis of the MAC unit has been targeted. This MAC unit is generally preferred for simpler designs. The experimental test shows that the results have been validated.

In this project, we propose a high speed low-power multiplier and accumulator (MAC) adopting the newSPST implementing approach. This MAC is designed by equipping the Spurious Power Suppression Technique (SPST) on a modified Booth encoder which is controlled by a detection unit using an AND gate. The modifiedbooth encoder will reduce the number of partial products generated by a factor of 2. The SPST adder will avoid the unwanted addition and thus minimize the switching power dissipation. The SPST MAC implementation with AND gates have an extremely high flexibility on adjusting the data asserting time. This facilitates the robustness of SPST can attain 30% speed improvement and 22% power reduction in the modified booth encoder. This design can be verified using Modelsim and Xilinx using verilog.

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The authors have no conflicts of interest to declare that are relevant to the content of this article.

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